

**UNIVERSITY OF SWAZILAND**  
**FACULTY OF SCIENCE & ENGINEERING**  
**DEPARTMENT OF ELECTRICAL & ELECTRONIC ENGINEERING**  
**SIGNALS AND SYSTEMS II**  
**COURSE CODE - EE332**  
**MAIN EXAMINATION**  
**MAY 2017**  
**DURATION OF THE EXAMINATION - 3 HOURS**

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**INSTRUCTIONS TO CANDIDATES**

1. There are **FIVE** questions in this paper. Answer any **FOUR** questions.
3. Show all your steps clearly in any calculations/work.
4. State clearly any assumptions made.
5. Start each new question on a fresh page.
6. Useful Fourier transform and Z-transform tables are **attached**.
7. Make sure that this exam contains 9 pages including this one.

**DO NOT OPEN THIS PAPER UNTIL PERMISSION HAS BEEN GIVEN BY THE INVIGILATOR.**

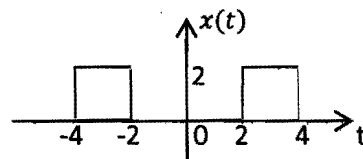
**QUESTION ONE (25 marks)**

(a) Suppose that  $x(t)$  has the Fourier transform  $u(2 - \omega^2)$ . Use the appropriate Fourier transform properties to find the Fourier transform of the signal  $x(4 - 3t)$ . [7]

(b) By using either calculation and/or Fourier transform table, find the Fourier transform  $X(\omega)$  of the following signals. [8+10]

(i)  $x(t) = e^{-3t} \cos(6t) u(t)$

(ii)



(Express your answer in terms of the sinc function.)

**QUESTION TWO (25 marks)**

(a) A particular LTI system is described by the difference equation

$$y(n) + \frac{1}{4}y(n - 1) - \frac{1}{8}y(n - 2) = x(n) - x(n - 1]$$

Determine the impulse response of the system. [6]

(b) Suppose that  $h(n) = \left(\frac{1}{2}\right)^n u(n)$ . Use discrete Fourier transforms to determine the response to the input signal  $x(n) = \left(\frac{3}{4}\right)^n u(n)$ . [6]

(c) Find the signal corresponding to the following Fourier transforms.

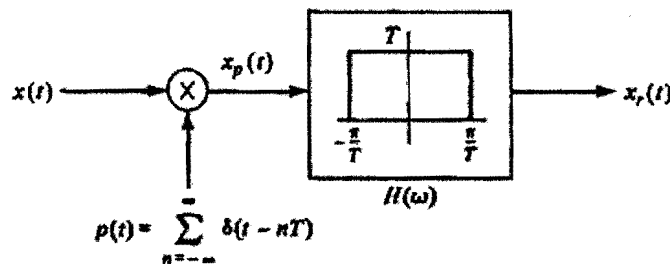
(i)  $X(e^{j\omega}) = \frac{2}{1 + \frac{1}{4}e^{-j(\omega - \frac{\pi}{2})}}$  [6]

(ii)  $X(e^{j\omega}) = \frac{16e^{j2\omega}}{1 - \frac{1}{4}e^{-j\omega}}$  [7]

**QUESTION THREE (25 marks)**

(a) State sampling theorem. Define Nyquist rate and Aliasing. [5]

(b) In the system shown in figure below,  $x(t)$  is sampled with a periodic impulse train, and a reconstructed signal  $x_r(t)$  is obtained from the samples by lowpass filtering.



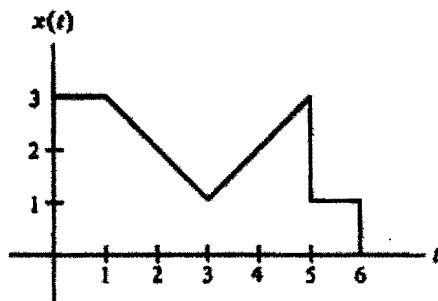
The sampling period  $T$  is  $1ms$ , and  $x(t)$  is a sinusoidal signal of the form  $x(t) = \cos(2\pi f_0 t + \theta)$ . For each of the following choices of  $f_0$  and  $\theta$ , determine  $x_r(t)$ . [20]

- (i)  $f_0 = 250Hz, \theta = \frac{\pi}{4}$
- (ii)  $f_0 = 750Hz, \theta = \frac{\pi}{2}$
- (iii)  $f_0 = 500Hz, \theta = \frac{\pi}{2}$

**QUESTION FOUR (25 marks)**

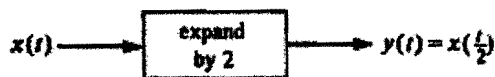
(a) What is amplitude modulation? Consider the signal  $x(t)$  shown in figure below.

[1+6+6]

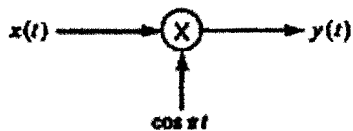


Draw  $y(t)$  for each of the following systems.

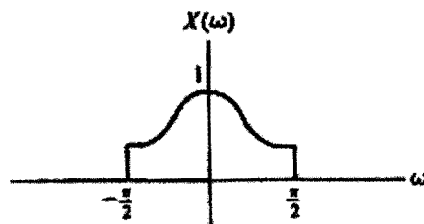
(i)



(ii)



(b) Suppose that  $x(t)$  has the Fourier Transform shown in figure below. Find  $Y(\omega)$  for each case in part (a). [6+6]



**QUESTION FIVE (25 marks)**

(a) Determine the z-transform (including the ROC) of the following sequences. Sketch the pole-zero plots and indicate the ROC on your sketch. [6]

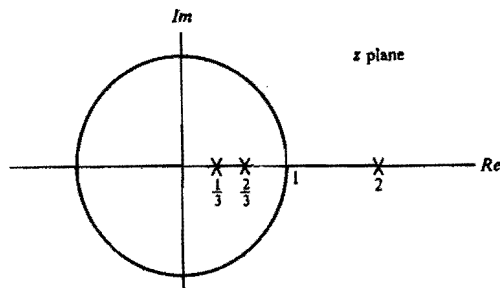
(i)  $\left(\frac{2}{3}\right)^n u(n)$

(ii)  $\delta(n + 2)$

(b) Find  $x(n)$  from  $X(z)$  below using partial fraction expansion, where  $x(n)$  is known to be causal. [7]

$$X(z) = \frac{3+2z^{-1}}{2+3z^{-1}+z^{-2}}$$

(c) Shown in figure below is the pole-zero plot for the z-transform  $X(z)$  of a sequence  $x(n)$ . [12]



Determine what can be inferred about the associated region of convergence from each of the following statement.

- (i)  $x(n)$  is right-sided.
- (ii)  $x(n)$  is left-sided.
- (iii) The Fourier transform of  $x(n)$  converges.

TABLE 1 Fourier Transforms

| No. | $x(t)$  | $X(\omega)$   |                             |
|-----|---|---|-----------------------------|
| 1   | $e^{-at}u(t)$   | $\frac{1}{a + j\omega}$   | $a > 0$                     |
| 2   | $e^{at}u(-t)$   | $\frac{1}{a - j\omega}$   | $a > 0$                     |
| 3   | $e^{-a t }$   | $\frac{2a}{a^2 + \omega^2}$                                     | $a > 0$                     |
| 4   | $te^{-at}u(t)$  | $\frac{1}{(a + j\omega)^2}$                                     | $a > 0$                     |
| 5   | $t^n e^{-at}u(t)$                                       | $\frac{n!}{(a + j\omega)^{n+1}}$                                | $a > 0$                     |
| 6   | $\delta(t)$   | 1   |                             |
| 7   | 1   | $2\pi\delta(\omega)$  |                             |
| 8   | $e^{j\omega_0 t}$                                       | $2\pi\delta(\omega - \omega_0)$                                 |                             |
| 9   | $\cos \omega_0 t$                                       | $\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$    |                             |
| 10  | $\sin \omega_0 t$                                       | $j\pi[\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$   |                             |
| 11  | $u(t)$  | $\pi\delta(\omega) + \frac{1}{j\omega}$                         |                             |
| 12  | $\text{sgn } t$   | $\frac{2}{j\omega}$   |                             |
| 15  | $e^{-at} \sin \omega_0 t u(t)$                          | $\frac{\omega_0}{(a + j\omega)^2 + \omega_0^2}$                 | $a > 0$                     |
| 16  | $e^{-at} \cos \omega_0 t u(t)$                          | $\frac{a + j\omega}{(a + j\omega)^2 + \omega_0^2}$              | $a > 0$                     |
| 17  | $\text{rect}\left(\frac{t}{\tau}\right)$                | $\tau \text{sinc}\left(\frac{\omega\tau}{2}\right)$             |                             |
| 18  | $\frac{W}{\pi} \text{sinc}(Wt)$                         | $\text{rect}\left(\frac{\omega}{2W}\right)$                     |                             |
| 19  | $\Delta\left(\frac{t}{\tau}\right)$                     | $\frac{\tau}{2} \text{sinc}^2\left(\frac{\omega\tau}{4}\right)$ |                             |
| 20  | $\frac{W}{2\pi} \text{sinc}^2\left(\frac{Wt}{2}\right)$ | $\Delta\left(\frac{\omega}{2W}\right)$                          |                             |
| 21  | $\sum_{n=-\infty}^{\infty} \delta(t - nT)$              | $\omega_0 \sum_{n=-\infty}^{\infty} \delta(\omega - n\omega_0)$ | $\omega_0 = \frac{2\pi}{T}$ |
| 22  | $e^{-t^2/2\sigma^2}$                                    | $\sigma\sqrt{2\pi}e^{-\sigma^2\omega^2/2}$                      |                             |

Table 2 Fourier Transform Operations

| Operation                             | $x(t)$                     | $X(\omega)$  |
|---------------------------------------|----------------------------|--|
| Scalar multiplication                 | $kx(t)$                    | $kX(\omega)$   |
| Addition                              | $x_1(t) + x_2(t)$          | $X_1(\omega) + X_2(\omega)$                          |
| Conjugation                           | $x^*(t)$                   | $X^*(-\omega)$                                       |
| Duality                               | $X(t)$                     | $2\pi x(-\omega)$                                    |
| Scaling ( $a$ real)                   | $x(at)$                    | $\frac{1}{ a } X\left(\frac{\omega}{a}\right)$       |
| Time shifting                         | $x(t - t_0)$               | $X(\omega)e^{-j\omega t_0}$                          |
| Frequency shifting ( $\omega_0$ real) | $x(t)e^{j\omega_0 t}$      | $X(\omega - \omega_0)$                               |
| Time convolution                      | $x_1(t) * x_2(t)$          | $X_1(\omega)X_2(\omega)$                             |
| Frequency convolution                 | $x_1(t)x_2(t)$             | $\frac{1}{2\pi} X_1(\omega) * X_2(\omega)$           |
| Time differentiation                  | $\frac{d^n x}{dt^n}$       | $(j\omega)^n X(\omega)$                              |
| Time integration                      | $\int_{-\infty}^t x(u) du$ | $\frac{X(\omega)}{j\omega} + \pi X(0)\delta(\omega)$ |

DISCRETE-TIME FOURIER TRANSFORM

A. Properties of the discrete-time Fourier transform

| Non-periodic signal  | Fourier transform   |
|--|---|
| $x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$   | $X(e^{j\omega}) \triangleq \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$   |
| $\left. \begin{matrix} x[n] \\ y[n] \end{matrix} \right\}$   | $\left. \begin{matrix} X(e^{j\omega}) \\ Y(e^{j\omega}) \end{matrix} \right\} \text{ Periodic with period } 2\pi$ |
| $ax[n] + by[n]$  | $aX(e^{j\omega}) + bY(e^{j\omega})$   |
| $x[n - n_0]$   | $e^{-j\omega n_0} X(e^{j\omega})$   |
| $e^{j\omega_0 n} x[n]$   | $X(e^{j(\omega - \omega_0)})$   |
| $x^*[n]$   | $X^*(e^{j(-\omega)})$   |
| $x[-n]$  | $X(e^{j(-\omega)})$   |
| $x_{(m)}[n] = \begin{cases} x[n/m], & n \text{ multiple of } m \\ 0, & n \text{ not multiple of } m \end{cases}$ | $X(e^{jm\omega})$   |
| $x[n] * y[n]$  | $X(e^{j\omega}) Y(e^{j\omega})$   |
| $x[n]y[n]$   | $\frac{1}{2\pi} \int_{2\pi} X(e^{j\theta}) Y(e^{j(\omega - \theta)}) d\theta$                                     |
| $x[n] - x[n - 1]$  | $(1 - e^{j\omega}) X(e^{j\omega})$  |
| $\sum_{k=-\infty}^n x[k]$  | $\frac{1}{1 - e^{j\omega}} X(e^{j\omega}) + \pi X(0) \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$           |
| $nx[n]$  | $j \frac{d}{d\omega} X(e^{j\omega})$  |

If  $x[n]$  is real valued then

$$x[n] \quad \left\{ \begin{array}{l} X(e^{j\omega}) = X^*(e^{j(-\omega)}) \\ \Re\{X(e^{j\omega})\} = \Re\{X(e^{j(-\omega)})\} \\ \Im\{X(e^{j\omega})\} = -\Im\{X(e^{j(-\omega)})\} \\ |X(e^{j\omega})| = |X(e^{j(-\omega)})| \\ \arg\{X(e^{j\omega})\} = -\arg\{X(e^{j(-\omega)})\} \end{array} \right.$$

$$\begin{array}{l} x_e[n] = \mathcal{E}\{x[n]\} \\ x_o[n] = \mathcal{O}\{x[n]\} \end{array} \quad \begin{array}{l} \Re\{X(e^{j\omega})\} \\ j\Im\{X(e^{j\omega})\} \end{array}$$

Parsevals relation for non-periodic signals

$$\sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{2\pi} |X(e^{j\omega})|^2 d\omega$$

B. Discrete-time Fourier transform table

| $x[n]$  | $X(e^{j\omega})$  |
|---|---|
| $\delta[n]$   | 1   |
| $\delta[n - n_0]$   | $e^{-j\omega n_0}$  |
| $\sum_{k=-\infty}^{\infty} \delta(n - kN)$  | $\frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta\left(\omega - \frac{2\pi k}{N}\right)$                             |
| 1   | $2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$  |
| $e^{j\omega_0 n}$   | $2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 - 2\pi k)$   |
| $\cos \omega_0 n$   | $\pi \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) + \delta(\omega + \omega_0 - 2\pi k)]$           |
| $\sin \omega_0 n$   | $\frac{\pi}{j} \sum_{k=-\infty}^{\infty} [\delta(\omega - \omega_0 - 2\pi k) - \delta(\omega + \omega_0 - 2\pi k)]$ |
| $u[n]$  | $\frac{1}{1 - e^{-j\omega}} + \pi \sum_{k=-\infty}^{\infty} \delta(\omega - 2\pi k)$                                |
| $a^n u(n),  a  < 1$   | $\frac{1}{1 - ae^{-j\omega}}$   |
| $(n+1)a^n u[n],  a  < 1$  | $\frac{1}{(1 - ae^{-j\omega})^2}$   |
| $\frac{(n+m-1)!}{n!(m-1)!} a^n u[n],  a  < 1$   | $\frac{1}{(1 - ae^{-j\omega})^m}$   |
| $\frac{1}{1 - a^2} a^{ n },  a  < 1$  | $\frac{1}{1 + a^2 - 2a \cos \omega}$  |
| $\begin{cases} 1, &  n  \leq N_1 \\ 0, & N_1 <  n  \leq \frac{N}{2} \end{cases}$<br>period $N$            | $2\pi \sum_{k=-\infty}^{\infty} a_k \delta\left(\omega - \frac{2\pi k}{N}\right)$                                   |
| $\begin{cases} 1, &  n  \leq N_1 \\ 0, &  n  > N_1 \end{cases}$   | $\frac{\sin \omega \left(N_1 + \frac{1}{2}\right)}{\sin \frac{\omega}{2}}$  |
| $\begin{cases} \frac{\sin Wn}{\pi} = \frac{W}{\pi} \text{sinc} \frac{Wn}{\pi} \\ 0 < W < \pi \end{cases}$ | $\begin{cases} 1, &  \omega  \leq W \\ 0, & W <  \omega  \leq \pi \end{cases}$<br>period $2\pi$                     |



## Table of Z-Transforms

| Line No. | $x(n), n \geq 0$       | z-Transform $X(z)$   | Region of Convergence |
|----------|------------------------|--|-----------------------|
| 1        | $x(n)$                 | $\sum_{n=0}^{\infty} x(n)z^{-n}$                                   |                       |
| 2        | $\delta(n)$            | 1  | $ z  > 0$             |
| 3        | $an(n)$                | $\frac{az}{z-1}$   | $ z  > 1$             |
| 4        | $na(n)$                | $\frac{z}{(z-1)^2}$  | $ z  > 1$             |
| 5        | $n^2a(n)$              | $\frac{z(z+1)}{(z-1)^3}$   | $ z  > 1$             |
| 6        | $a^n x(n)$             | $\frac{z}{z-a}$  | $ z  >  a $           |
| 7        | $e^{-an} x(n)$         | $\frac{z}{(z-e^{-a})}$   | $ z  > e^{-a}$        |
| 8        | $na^n x(n)$            | $\frac{az}{(z-a)^2}$   | $ z  >  a $           |
| 9        | $\sin(an)x(n)$         | $\frac{z \sin(a)}{z^2 - 2z \cos(a) + 1}$                           | $ z  > 1$             |
| 10       | $\cos(an)x(n)$         | $\frac{z[z - \cos(a)]}{z^2 - 2z \cos(a) + 1}$                      | $ z  > 1$             |
| 11       | $a^n \sin(bn)x(n)$     | $\frac{[a \sin(b)]z}{z^2 - [2a \cos(b)]z + a^2}$                   | $ z  >  a $           |
| 12       | $a^n \cos(bn)x(n)$     | $\frac{z[z - a \cos(b)]}{z^2 - [2a \cos(b)]z + a^2}$               | $ z  >  a $           |
| 13       | $e^{-an} \sin(bn)x(n)$ | $\frac{[e^{-a} \sin(b)]z}{z^2 - [2e^{-a} \cos(b)]z + e^{-2a}}$     | $ z  > e^{-a}$        |
| 14       | $e^{-an} \cos(bn)x(n)$ | $\frac{z[z - e^{-a} \cos(b)]}{z^2 - [2e^{-a} \cos(b)]z + e^{-2a}}$ | $ z  > e^{-a}$        |

### Properties of Z-Transforms

Linearity:  $ax_1[k] + bx_2[k] \Leftrightarrow aX_1(z) + bX_2(z)$

Time Reversal:  $x[-k] \Leftrightarrow X(1/z)$

Summation:  $\sum_{n=-\infty}^k x[n] \Leftrightarrow \frac{zX(z)}{z-1}$

Initial Value:  $x[0] = \lim_{z \rightarrow \infty} X(z)$

Final Value:  $x[\infty] = \lim_{z \rightarrow 1} (z-1)X(z)$

Convolution:  $x[k] * h[k] \Leftrightarrow X(z)H(z)$

Differencing:  $x[k] - x[k-1] \Leftrightarrow (1-z^{-1})X(z)$

Differentiation:  $-kx[k] \Leftrightarrow z \frac{d}{dz} X(z)$

Time Shifting:  $x[n-n_0] \Leftrightarrow z^{-n_0} X(z), n_0 \geq 0$

$$x[n+n_0] \Leftrightarrow z^{n_0} \left( X(z) - \sum_{m=0}^{n_0-1} x[m]z^{-m} \right), n_0 \geq 0$$