UNIVERSITY OF SWAZILAND

FACULTY OF SCIENCE AND ENGINEERING

DEPARTMENT OF PHYSICS

MAIN EXAMINATION 2013/2014

TITLE OF PAPER : ELECTROMAGNETIC THEORY

COURSE NUMBER : P331

TIME ALLOWED : THREE HOURS

INSTRUCTIONS : ANSWER <u>ANY FOUR</u> OUT OF FIVE QUESTIONS. EACH QUESTION CARRIES <u>25</u> MARKS.

MARKS FOR DIFFERENT SECTIONS ARE SHOWN IN THE RIGHT-HAND MARGIN.

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THIS PAPER HAS <u>NINE</u> PAGES, INCLUDING THIS PAGE.

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P331 ELECTROMAGNETIC THEORY

Question one

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A uniform surface charge density of $\left(\frac{k}{a}\right)C/m^2$ where k is a positive constant is deposited on $\rho = a$ surface of a very long straight coaxial cable, with an inner solid wire of radius a m. A uniform surface charge density of $\left(-\frac{k}{c}\right)C/m^2$ is deposited on $\rho = c$ surface of the outer hollow wire with inner radius of c m and outer radius of d m. In-between the wires is filled with two layers of insulating materials with permittivity $\varepsilon_1 \& \varepsilon_2$ for $(a < \rho < b) \& (b < \rho < c)$ layers respectively as shown in the figure below.



- (a) Find the total electric charges per unit length deposited on each of the inner and outer conducting wires in terms of k and name them as $q_1 \& q_2$ respectively. Show that $q_1 = -q_2$. (3 marks)
- (b) Set $\vec{D}(\rho, \phi, z) = \vec{e}_{\rho} D_{\rho}(\rho)$ (write a brief justification for this setting), draw an appropriate closed surface and utilize the integral form of electric Gauss law, i.e., $\oint \vec{D} \cdot d\vec{s} = \{ total \ Q \ enclosed \ by \ S \}$, to find $D_{\rho}(\rho)$ for $a < \rho < c$ region. Then

write down the form of electric field \vec{E} for $a < \rho < c$ region. (2+2+6 marks) (c) Find the potential difference between the inner and outer wires and then write down

(d) the distributive capacitance of the given coaxial cable system. (5+2 marks)(d) Explain briefly the bounded charge excess on the surface of a dielectric material and then find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the find the surface charge density of the bounded charge excess σ_{sp} on the surface charge d

 $\rho = a \& \rho = c$ dielectric material surfaces in terms of k, $\varepsilon_1 \& \varepsilon_2$. (1+4 marks) (Hint: $\sigma_{sp} = \vec{e}_n \bullet \vec{P}$ where \vec{e}_n is the normal outward unit vector on dielectric surface)

Question two

(a) A thin conducting wire of length 2L, with its central axis coinciding with the z axis and its centre point coinciding with the origin, carries a steady total current I along positive z direction as shown in the figure below.



Since the given current source is only along z axis, its produced vector potential at a field point $P(\rho, \phi, 0)$ is also having only z component A_z , i.e.,

$$\vec{A} = \vec{e}_z A_z \quad \text{where} \quad A_z = \int_{z'=-L}^{z'=+L} \frac{\mu_0 \ I \ d \ z'}{4 \ \pi \ R} = 2 \int_{z'=0}^{z'=+L} \frac{\mu_0 \ I \ d \ z'}{4 \ \pi \ \sqrt{(z')^2 + \rho^2}} ,$$

carry out the above integral for A_z about z' and show that

$$A_{z} = \frac{\mu_{0} I}{2 \pi} \ln \left(\frac{L + \sqrt{L^{2} + \rho^{2}}}{\rho} \right)$$
 (9 marks)

(Hint: set $z' = \rho \tan(\alpha)$, $\int \sec(\alpha) d\alpha = \ln(\sec(\alpha) + \tan(\alpha))$)

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$$L >> \rho$$
, *i.e.*, $\sqrt{L^2 + \rho^2} \rightarrow L$, use $\vec{A} \approx \vec{e}_z \frac{\mu_0 I}{2\pi} \ln\left(\frac{2L}{\rho}\right)$ and $\vec{B} = \vec{\nabla} \times \vec{A}$

to find the magnetic field \vec{B} at the field point and show that

$$=\vec{e}_{\phi}\frac{\mu_0 I}{2\pi\rho} \quad .$$

(6 marks)

Question two (continued)

(b) Placing a rectangular conducting loop of dimension $b \times c$ a distance of d away from the central current carrying wire as shown in the diagram below, i.e., the inner region confined by the rectangular loop in clockwise sense is

 $S: d \le \rho \le d + b \quad , \quad 0 \le z \le c \quad \& \quad d \, \vec{s} = \vec{a}_{\phi} \, d \, \rho \, d \, z \ ,$



(i)

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- find the total magnetic flux Φ_m passing through the inner region confined by the rectangular loop, i.e., $\Phi_m = \int_{\mathcal{S}} \vec{B} \cdot d\vec{s}$, in terms of μ_0 , b, c, d & I. Also write down the mutual inductance M between the given rectangular loop and the long straight wire and show that $M = \frac{\mu_0}{2\pi} \times \ln\left(\frac{d+b}{d}\right) \times c$. (5 marks)
- (ii) If given the values of $b=5\ cm$, $c=8\ cm$ & $d=10\ cm$ find the value of M. Further if the wire carries a sinusoidal current $I(t)=2\sin(9t)$ A instead of carrying a static current I, find the induced e.m.f. in the rectangular conducting loop. $(3+2\ marks)$

Question three

(a) An interface separating two dielectric regions of permittivity $\varepsilon_1 \& \varepsilon_2$ is shown in the figure below. $\vec{E}_1 \& \vec{E}_2$ are the electric fields at the same point on the interface in the different regions and $\theta_1 \& \theta_2$ are their respective angles made with the normal.



- (i) Use the integral Faraday's law and by choosing proper closed loop across the interface, deduce that the tangential component of \vec{E} is continuous at the interface, i.e., $E_{1t} = E_{2t}$. (5 marks)
- (ii) Use the integral electric Gauss law and by choosing proper closed surface across the interface, deduce that the normal component of \vec{D} is continuous at the interface, i.e., $D_{1n} = D_{2n}$. (5 marks)
- (iii) Use the results in (a)(i) and (a)(ii) and deduce the following refraction law for \vec{E} $\varepsilon_1 \tan(\theta_2) = \varepsilon_2 \tan(\theta_1)$. (4 marks)
- (b) The equation of motion of an average conduction electron based on Drude's model is $m_e \frac{d\vec{v}_d}{dt} = -e \vec{E} \frac{2 m_e \vec{v}_d}{\tau_c} ,$
 - (i) Explain briefly each term in the above equation. (2 marks)
 - (ii) In the steady state situation, i.e., $\frac{d\vec{v}_d}{dt} = 0$, deduce the following point form of Ohm's law $\vec{J} = \sigma \vec{E}$ where $\sigma = \frac{ne^2}{2m_e}\tau_c$, (Hint : $\vec{J} = \rho_v \vec{v}_d = -ne\vec{v}_d$) (5 marks)
 - (iii) If a certain pure metal has an atomic density of $3 \neq 10^{28} \frac{atoms}{m^3}$ at room temperature and three outer orbit electrons are conduction electrons, find the value of τ_c if its measured dc conductivity is $\sigma = 2 \times 10^7 \frac{1}{m \Omega}$. (4 marks)

Question four

140 The Maxwell's equations for the material region of parameters μ , ε & σ (a) are $\vec{\nabla}\bullet\vec{E}=0$ (1) $\vec{\nabla} \bullet \vec{H} = 0$ (2) $\vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$ (3) $\vec{\nabla} \times \vec{H} = \sigma \, \vec{E} + \varepsilon \, \frac{\partial \vec{E}}{\partial t}$ (4) Deduce from them the following wave equation for \vec{H} (i) $\nabla^2 \vec{H} - \mu \sigma \frac{\partial \vec{H}}{\partial t} - \mu \varepsilon \frac{\partial^2 \vec{H}}{\partial t^2} = 0 \quad \dots \quad (5) \quad .$ (3 marks) Set \vec{H} as $\left(\vec{H}(space)e^{i\omega t}\right)$ and substitute it into the above wave equation, (ii) to deduce the following time-harmonic equation for $\hat{H}(space)$ as $\nabla^2 \vec{H}(space) - \hat{\gamma}^2 \vec{H}(space) = 0$ where $\hat{\gamma} = \sqrt{i \omega \mu \sigma - \omega^2 \mu \varepsilon}$ (3 marks) Set the propagation constant $\hat{\gamma} \equiv \alpha + i \beta$, to deduce that (iii) $\beta = \frac{\omega \sqrt{\mu \varepsilon}}{\sqrt{2}} \sqrt{\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^2} + 1}$ (6 marks) (Hint: $\sin\left(\frac{\theta}{2}\right) = \sqrt{\frac{1-\cos(\theta)}{2}} \& \cos(\theta) = \left(\sqrt{1+\tan^2(\theta)}\right)^{-1}$) An uniform plane wave traveling along the +z direction with the field components (b) $E_x(z) \& H_y(z)$ has a complex electric field amplitude $\hat{E}_m = 100 e^{i 20^\circ} \frac{V}{m}$ and propagates at a frequency $f = 10^6$ Hz in a material region having the parameters $\mu = \mu_0$, $\varepsilon = 2 \varepsilon_0$ & $\frac{\sigma}{\omega c} = 0.4$. Find the values of the propagation constant $\hat{\gamma} (= \alpha + i \beta)$ and the intrinsic wave (i) impedance $\hat{\eta}$ for this wave, (4 marks) Express the electric and magnetic fields in both their complex and real-time forms, (ii) with the numerical values of (b)(i) inserted, (6 marks) Find the values of the penetration depth, wavelength and phase velocity of the given (iii) wave. (3 marks) { 6

Question five

"An uniform plane wave $(\hat{E}_{x1}^+, \hat{H}_{y1}^+)$, operating at a frequency f, is normally incident upon a layer of d_2 thickness, and emerges to region 3 as shown below :

 $O_1\,$, $O_2\,$ & $O_3\,$ are the respective origins for region 1 , 2 & 3 chosen at the first and second interface .

(a) Define for the *i*th region (*i* = 1,2,3) the reflection coefficient $\hat{\Gamma}_i(z)$ and the total wave impedance $\hat{Z}_i(z)$ and deduce the following : $\hat{\Gamma}(z) = \hat{Z}_i(z) - \hat{\eta}_i$ (6 more the following)

$$\hat{\Gamma}_i(z) = \frac{Z_i(z) - \hat{\eta}_i}{\hat{Z}_i(z) - \hat{\eta}_i}$$
(6 marks)

(b) If $f = 10^8$ Hz & $d_2 = \frac{\lambda_2}{4}$, region 1 & 3 are air regions and region 2 is a lossless

region with parameters $\mu_2 = \mu_0$, $\varepsilon_1 = 4 \varepsilon_0$ & $\frac{\sigma}{\omega \varepsilon} = 0$

- (i) find the values of β_1 , β_2 , β_3 , λ_2 & $\hat{\eta}_2$. (note: $\hat{\eta}_1 = \hat{\eta}_3 = 120 \pi \Omega$ and $\alpha_1 = \alpha_2 = \alpha_3 = 0$) (4 marks)
- (ii) Starting with $\hat{\Gamma}_3(z) = 0$ for the rightmost region, i.e., region 3, and using continuous \hat{Z} at the interface as well as the equations in (a), find the values of $\hat{Z}_3(0)$, $\hat{Z}_2(0)$, $\hat{\Gamma}_2(0)$, $\hat{\Gamma}_2(-d_2)$, $\hat{Z}_2(-d_2)$, $\hat{Z}_1(0)$ & $\hat{\Gamma}_1(0)$ (10 marks)
- (iii) Find the value of \hat{E}_{m1}^- & \hat{E}_{m2}^+ if given $\hat{E}_{m1}^+ = 100 e^{i^{30^0}} \frac{V}{m}$ (5 marks)

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Useful informations

$$e = 1.6 \times 10^{-19} C$$

$$m_{e} = 9.1 \times 10^{-31} kg$$

$$\mu_{0} = 4 \pi \times 10^{-7} \frac{H}{m}$$

$$\varepsilon_{0} = 8.85 \times 10^{-12} \frac{F}{m}$$

$$\alpha = \frac{\omega \sqrt{\mu \varepsilon}}{\sqrt{2}} \sqrt{\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^{2} - 1}}$$

$$\beta = \frac{\omega \sqrt{\mu \varepsilon}}{\sqrt{2}} \sqrt{\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^{2} + 1}}$$

$$\frac{1}{\sqrt{\mu_{0}} \varepsilon_{0}} = 3 \times 10^{8} \frac{m}{s}$$

$$\hat{\eta} = \frac{\sqrt{\frac{\mu}{\varepsilon}}}{\sqrt{1 + \left(\frac{\sigma}{\omega \varepsilon}\right)^{2}}} e^{i\frac{1}{2}un^{-1}\left(\frac{\sigma}{\omega \varepsilon}\right)}$$

$$\eta_{0} = \sqrt{\frac{\mu_{0}}{\varepsilon_{0}}} = 120 \pi \Omega = 377 \Omega$$

$$\beta_{0} = \omega \sqrt{\mu_{0}} \varepsilon_{0}$$

$$\iint_{s} \vec{E} \cdot d\vec{s} = \frac{1}{\varepsilon} \iiint_{v} \rho_{v} dv$$

$$\oiint_{s} \vec{B} \cdot d\vec{s} = 0$$

$$\iint_{L} \vec{E} \cdot d\vec{l} = -\frac{\partial}{\partial t} (\iint_{s} \vec{B} \cdot d\vec{s})$$

$$\vec{\nabla} \cdot \vec{E} = \frac{\rho_{v}}{\varepsilon}$$

$$\vec{\nabla} \cdot \vec{B} = 0$$

$$\vec{\nabla} \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$

$$\vec{J} = \sigma \vec{E}$$

$$\begin{split} \vec{D} &= c \ \vec{E} = c_0 \ \vec{E} + \vec{P} \qquad \& \qquad \vec{B} = \mu \ \vec{H} = \mu_0 \ \vec{H} + \vec{M} \\ & \oint_{I_s} \vec{F} \cdot d\vec{s} &= \oint_{I_s} (\vec{\nabla} \times \vec{F}) dv \qquad \text{divergence theorem} \\ & \oint_{L} \vec{F} \cdot d\vec{s} &= \iint_{S} (\vec{\nabla} \times \vec{F}) \cdot d\vec{s} \qquad \text{Stokes' theorem} \\ & \vec{\nabla} \cdot (\vec{\nabla} \times \vec{F}) &= 0 \\ & \vec{\nabla} \times (\vec{\nabla} \times \vec{F}) &= 0 \\ & \vec{\nabla} \times (\vec{\nabla} \times \vec{F}) &= \vec{\nabla} (\vec{\nabla} \cdot \vec{F}) - \nabla^2 \ \vec{F} \\ & \vec{\nabla} f &= \vec{e}_s \ \frac{\partial f}{\partial x} + \vec{e}_s \ \frac{\partial f}{\partial y} + \vec{e}_s \ \frac{\partial f}{\partial z} &= \vec{e}_\rho \ \frac{\partial f}{\partial \rho} + \vec{e}_s \ \frac{1}{\rho} \ \frac{\partial f}{\partial \rho} + \vec{e}_s \ \frac{\partial f}{\partial z} \\ &= \vec{e}_\rho \ \frac{\partial f}{\partial x} + \vec{e}_\rho \ \frac{1}{\rho} \ \frac{\partial f}{\partial \theta} + \vec{e}_s \ \frac{1}{100} \ \frac{\partial f}{\partial \phi} \\ & \vec{\nabla} \cdot \vec{F} &= \frac{\partial (F_s)}{\partial x} + \frac{\partial (F_s)}{\partial y} + \frac{\partial (F_s)}{\partial z} \\ &= \frac{1}{\rho} \ \frac{\partial (F_s \ r)}{\partial x} + \frac{\partial (F_s)}{\partial z} + \frac{1}{\rho} \ \frac{\partial (F_s \ sin(\theta))}{\partial \theta} + \frac{1}{\rho} \ \frac{\partial (F_s)}{\partial \phi} + \frac{\partial (F_s)}{\partial z} \\ &= \frac{1}{r^2} \ \frac{\partial (F_s \ r)}{\partial x} + \frac{\partial (F_s)}{\partial z} \\ &= \frac{1}{r^2} \ \frac{\partial (F_s \ r)}{\partial \phi} - \frac{\partial (F_s)}{\partial z} \\ &= \frac{\vec{e}_s}{\rho} \left(\frac{\partial (F_s)}{\partial \phi} - \frac{\partial (F_s)}{\partial z} \right) \\ &= \frac{\vec{e}_s}{\rho} \left(\frac{\partial (F_s \ r)}{\partial \phi} - \frac{\partial (F_s \ r)}{\partial z} \right) \\ &= \frac{\vec{e}_s}{\rho} \left(\frac{\partial (F_s \ r)}{\partial \phi} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta))}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s}{r^2} \sin(\theta) \left(\frac{\partial (F_s \ rin(\theta)}{\partial \theta} - \frac{\partial (F_s \ r)}{\partial \phi} \right) \\ &= \frac{\vec{e}_s \ dx \ e \vec{e}_s \ r \ e \vec{e}_s \ F_s \ e$$

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